# Linear Filters in StreamIt 

Andrew A. Lamb MIT
Laboratory for Computer Science
Computer Architecture Group
8/29/2002

## Outline

- Introduction
- Dataflow Analysis
- Hierarchal Matrix Combinations
- Performance Optimizations


## Basic Idea



## What is a Linear Filter?

- Generic filters calculate some outputs (possibly) based on their inputs.
- Linear filters: outputs $\left(y_{j}\right)$ are weighted sums of the inputs ( $\mathrm{x}_{\mathrm{i}}$ ) plus a constant.

$$
\begin{aligned}
& y=\sum_{i \in[1, N]} w_{i} x_{i}+b \quad \begin{array}{l}
\text { for } \mathrm{b} \text { constant } \\
\mathrm{w}_{\mathrm{i}} \text { constant for all i } \\
\mathrm{N} \text { is the number of inputs }
\end{array} \\
& y=w_{1} x_{1}+w_{2} x_{2}+(\ldots)+w_{N} x_{N}+b
\end{aligned}
$$

## Linearity and Matricies

- Matrix multiply is exactly weighted sum
- We treat inputs $\left(\mathrm{x}_{\mathrm{i}}\right)$ and outputs $\left(\mathrm{y}_{\mathrm{j}}\right)$ as vectors of values ( $\mathbf{x}$, and $\mathbf{y}$ respectively)
- Filter is represented as a matrix of weights $\mathbf{A}$ and a vector of constants $\mathbf{b}$
- Therefore, filter represents the equation $\mathbf{y}=\mathbf{x A}+\mathbf{b}$


## Equation Example, $\mathrm{y}_{1}$

$$
\left.\left[\begin{array}{llll}
x_{1} & x_{2} & x_{3} & x_{4}
\end{array}\right]\left[\begin{array}{lll}
a_{1,} \\
a_{2,1} & a_{1,2} & a_{1,3} \\
a_{3,1} \\
a_{2,2} & a_{2,3} \\
a_{4} & a_{3,2} & a_{3,3} \\
a_{4,2} & a_{4,3}
\end{array}\right]+\left(\begin{array}{ll}
\left.b_{b}\right) & b_{2} \\
b_{3}
\end{array}\right]+\begin{array}{lll}
y_{1} & y_{2} & y_{3}
\end{array}\right]
$$

$$
y_{1}=\left(x_{1} a_{1,1}+x a_{2,1}+x_{3} a_{3,1}+x_{4} a_{4,1}+b_{1}\right)
$$

## Equation Example, $y_{2}$

$$
\left[\begin{array}{llll}
x_{1} & x_{2} & x_{3} & x_{4}
\end{array}\right]\left[\begin{array}{lll}
a_{1,1} & a_{1,2} & a_{1,3} \\
a_{2,1} & a_{2,2} & a_{2,3} \\
a_{3,1} & a_{3,2} & a_{3,3} \\
a_{4,1} & a_{4} & a_{4,3}
\end{array}\right]+\left[\begin{array}{lll}
b_{1} & b_{2} & b_{3}
\end{array}\right]+\left[\begin{array}{lll}
y_{1} & y_{2} & y_{3}
\end{array}\right]
$$

$$
y_{2}=\left(x_{1} a_{1,2}+x a_{2,2}+x_{3} a_{3,2}+x_{4} a_{4,2}+b_{2}\right)
$$

## Equation Example, $y_{3}$

$$
\left[\begin{array}{llll}
x_{1} & x_{2} & x_{3} & x_{4}
\end{array}\right]\left[\begin{array}{lll}
a_{1,1} & a_{1,2} & a_{1,1} \\
a_{2,1} & a_{2,2} & a_{2,3} \\
a_{3,1} & a_{3,2} & a_{3,3} \\
a_{4,1} & a_{4,2} & a_{4,3}
\end{array}\right]+\left[\begin{array}{lll}
b_{1} & b_{2} & b_{3}
\end{array}\right]+\left[\begin{array}{lll}
y_{1} & y_{2} & y_{3}
\end{array}\right]
$$

$$
y_{3}=\left(x_{1} a_{1,3}+x a_{2,3}+x_{3} a_{3,3}+x_{4} a_{4,3}+b_{3}\right)
$$

## Usefulness of Linearity

- Not all filters compute linear functions
- push(pop()*pop());
- Many fundamental DSP filters do
- DFT/FFT
- DCT
- Convolution/FIR
- Matrix Multiply


## Example: DFT Matrix

$$
\begin{aligned}
& \text { DFT: } X(m) \equiv \sum_{n=0}^{N-1} x(n) e^{-j 2 \pi n m / N}, m=0,1,2, \ldots, N-1 \\
& F_{N}=\left[\begin{array}{cccc}
w_{N}^{0 \bullet 0} & w_{N}^{0 \bullet 1} & \cdots & w_{N}^{0 \bullet(N-1)} \\
w_{N}^{1 \bullet 0} & w_{N}^{1 \bullet 1} & \cdots & w_{N}^{1 \bullet(N-1)} \\
\cdots & \ldots & \cdots & \cdots \\
w_{N}^{(N-1) \bullet 0} & w_{N}^{(N-1) \bullet 1} & \ldots & w_{N}^{(N-1) \bullet(N-1)}
\end{array}\right] \begin{array}{l}
\text { row n } \\
\text { column m }
\end{array} \\
& w_{N}=e^{-j 2 \pi / N}
\end{aligned}
$$

## Example: IDFT Matrix

IDFT: $x(n) \equiv \frac{1}{N} \sum_{m=0}^{N-1} X(m) e^{j 2 \pi m n / N}, n=0,1,2, \ldots, N-1$

$$
\begin{aligned}
& F_{N}^{-1}=\left[\begin{array}{cccc}
w_{N}^{0 \bullet 0} & w_{N}^{0 \bullet 1} & \cdots & w_{N}^{0 \bullet(N-1)} \\
w_{N}^{1 \bullet 0} & w_{N}^{1 \bullet 1} & \cdots & w_{N}^{1 \bullet(N-1)} \\
\cdots & \cdots & \cdots & \cdots \\
w_{N}^{(N-1) \bullet 0} & w_{N}^{(N-1) \bullet 1} & \cdots & w_{N}^{(N-1) \bullet(N-1)}
\end{array}\right] \begin{array}{c}
\text { row m } \\
\text { column } \mathrm{n}
\end{array} \\
& w_{N}=\left(\frac{1}{N}\right) e^{j 2 \pi / N}
\end{aligned}
$$

## Usefullness, cont.

- Matrix representations
- Are "embarrassingly parallel" 1
- Expose redundant computation
- Let us take advantage of existing work in DSP field
- Well understood mathematics


## Outline

- Introduction
- Dataflow Analysis
- Hierarchal Matrix Combinations
- Performance Optimizations


## Dataflow Analysis

- Basic idea: convert the general code of a filter's work function into an affine representation (eg $\mathbf{y = x} \mathbf{A}+\mathbf{b}$ )
- The A matrix represents the linear combination of inputs used to calculate each output.
- The vector b represents a constant offset that is added to the combination.


## "Linear" Dataflow Analysis

- Much like standard constant prop.
- Goal: Have a vector of weights and a constant that represents the argument to each push statement which become a column in $\mathbf{A}$ and an entry in $\mathbf{b}$.
- Keep mappings from variables to their linear forms (eg vector + constant).


## "Linear" Dataflow Analysis

- Of course, we need the appropriate generating cases, eg
- constants $\longrightarrow\left[\begin{array}{l}0 \\ 0 \\ 0\end{array}\right]+c$
- pop/peek $(x) \longrightarrow\left[\begin{array}{l}0 \\ 1 \\ 0\end{array}\right]+0$


## "Linear" Dataflow Analysis

- Like const prop, confluence operator is set union.
- Need combination rules to handle things like multiplication and addition (vector add and constant scale)


## Ridiculous Example

```
a=peek (2);
b=pop();
c=pop ();
pop();
d=a+2b;
e=d+5;
```


## Ridiculous Example

```
a=peek (2);
b=pop ();
c=pop ();
pop () ;
d=a+2b;
e=d+5;
```



## Ridiculous Example

```
a=peek (2);
b=pop ();
c=pop();
pop();
d=a+2b;
e=d+5;
```



## Ridiculous Example

```
a=peek (2);
b=pop();
c=pop ();
pop();
d=a+2b;
e=d+5;
```



## Ridiculous Example

```
a=peek (2);
b=pop();
c=pop();
pop();
d=a+2b;
e=d+5;
```



## Ridiculous Example

```
a=peek (2);
b=pop();
c=pop();
pop();
d=a+2b;
e=d+5;
```



## Constructing matrix $\mathbf{A}$

Filter A:
filter code
push (b);
push (a);
$\mathbf{a} \longrightarrow\left[\begin{array}{l}1 \\ 2 \\ 3\end{array}\right]+4$
$\mathbf{b} \longrightarrow\left[\begin{array}{l}5 \\ 6 \\ 7\end{array}\right]+8$

$$
A=\left[\begin{array}{ll}
0 & 0 \\
0 & 0 \\
0 & 0
\end{array}\right] \quad b=\left[\begin{array}{ll}
0 & 0
\end{array}\right]
$$

## Constructing matrix $\mathbf{A}$

Filter A:
filter code
push (b); $\longleftarrow$ push (a);
$\mathbf{a} \longrightarrow\left[\begin{array}{l}1 \\ 2 \\ 3\end{array}\right]+4$
$\mathbf{b} \longrightarrow\left[\begin{array}{l}5 \\ 6 \\ 7\end{array}\right]+8$

$$
A=\left[\begin{array}{ll}
0 & 5 \\
0 & 6 \\
0 & 7
\end{array}\right] \quad b=\left[\begin{array}{ll}
0 & 8
\end{array}\right]
$$

## Constructing matrix $\mathbf{A}$

Filter A:
push (b);
$\mathbf{a} \longrightarrow\left[\begin{array}{l}1 \\ 2 \\ 3\end{array}\right]+4$
$\mathbf{b} \longrightarrow\left[\begin{array}{l}5 \\ 6 \\ 7\end{array}\right]+8$
$A=\left[\begin{array}{ll}1 & 5 \\ 2 & 6 \\ 3 & 7\end{array}\right] \quad b=\left[\begin{array}{ll}4 & 8\end{array}\right]$

## Big Picture



## Big Picture, cont.



## Outline

- Introduction
- Dataflow Analysis
- Hierarchal Matrix Combinations
- Performance Optimizations


## Combining Filters

- Basic idea: combine pipelines, splitjoins (and possibly feedback loops) of linear filters together
- End up with a single large matrix representation
- The details are tricky in the general case (eg I am still working on them)


## Combining Pipelines


ye olde pipeline


The matrix $\mathbf{C}$ is
calculated as $\mathbf{A}^{\prime} \mathbf{B}^{\prime}$ where $\mathbf{A}^{\prime}$ and $\mathbf{B}^{\prime}$ have been appropriately scaled and duplicated to make the dimensions work out.

In the case where peek(B) $=$ pop(B), we might have to use two stage filters or duplicate some work to get the dimensions to work out.

## Combining Split Joins



A split join reorders data, so the columns of $\mathbf{C}$ are interleaved copies of the columns of $\mathrm{A}_{1}$ through $\mathrm{A}_{\mathrm{N}}$.

Matching the rates of $A_{1}$ through $A_{N}$ is a challenge that I am still working out.

## Combining Feedback Loops


ye olde FeedbackLoop
It is unclear if we can do anything of use with a FeedbackLoop construct. Eigen values might give information about stability, but it is not clear if that is useful... more thought is needed.

## Outline

- Introduction
- Dataflow Analysis
- Hierarchal Matrix Combinations
- Performance Optimizations


## Performance Optimizations

- Take advantage of our compile time knowledge of the matrix coefficients.
- eg don't waste computation on zeros
- Try and leverage existing DSP work on factoring matricies.
- Try to recognize parallel structures in our matrices.
- Use frequency analysis.


## Factoring for Performance

$$
A=\left[\begin{array}{llll}
a_{1,1} & a_{1,2} & a_{1,3} & a_{1,4} \\
a_{2,1} & a_{2,2} & a_{2,3} & a_{2,4} \\
a_{3,1} & a_{3,2} & a_{3,3} & a_{3,4} \\
a_{4,1} & a_{4,2} & a_{4,3} & a_{4,4}
\end{array}\right]
$$

16 multiplies
12 adds

$$
B C=\left[\begin{array}{cccc}
b_{1,1} & 0 & 0 & 0 \\
0 & b_{2,2} & 0 & 0 \\
0 & 0 & b_{3,3} & 0 \\
0 & 0 & 0 & b_{4,4}
\end{array}\right]\left[\begin{array}{cccc}
c_{1,1} & 0 & 0 & 0 \\
c_{2,1} & c_{2,2} & 0 & 0 \\
c_{3,1} & c_{3,2} & c_{3,3} & 0 \\
c_{4,1} & c_{4,2} & c_{4,3} & c_{4,4}
\end{array}\right]
$$

14 multiplies
6 adds

## SPL/SPIRAL

- Software package that will attempt to find a fast implementation signal processing algorithms described as matrices.
- It attempts to find a sparse factorization of an arbitrary matrix.
- It can automatically derive FFT (eg the Cooley-Turkey algorithm) from DFT definition.
- Claim that their performance is $\approx$ FFTW $^{1}$.


## Recognize Parallel Structure

- We can go from SplitJ oin to matrix.
- Perhaps we can recognize the reverse transformation.
- Also, implement blocked matrix multiply to keep parallel resources busy.


## Frequency Analysis

Instead of computing the matrix product straight up, possibly go to frequency domain.


Rids us of offset vector (added to response at $\mathrm{f}=0$ ).
Might allow additional optimizations (because of possible symmetries exposed in frequency domain).


## Work left to do

- Implementation of single filter analysis.
- Combining hierarchical constructs.
- Understand the math of automatic matrix factorizations (group theory).
- Analyze frequency analysis.
- Implement optimizations.
- Get results.


## Questions for the Future

- Are there any other optimizations?
- Can we produce inverted matrices
- programmer codes up transmitter and StreamIt automatically creates the receiver. ${ }^{1}$
- How many cycles of a "real" DSP application are spent computing linear functions?
- Can we combine the linear description of what happens inside a filter with the SARE representation of what is happening between them? (POPL paper)

